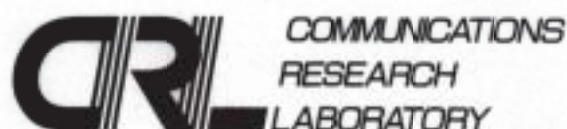


INV 18.7405 (100)
NAND 18.7400 (340)

林 正彦 様



1992年 2月 25日

この度の観測では大変お世話になりました。ありがとうございました。観測も無事終了し、おかげさまでたくさんのデータも取ることができました。みな様の貴重な観測の時間をさいっていただき、感謝しております。どうぞみな様によりしくお伝えいたしますようお願いいたします。長谷川先生には別にお電話でお礼申し上げますか、よろしくお伝え下さい。

増子 治信

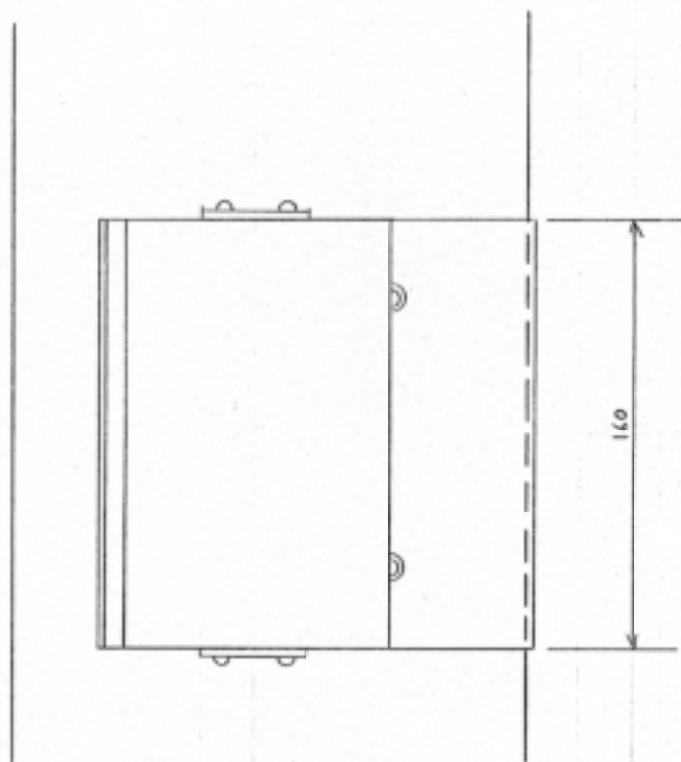
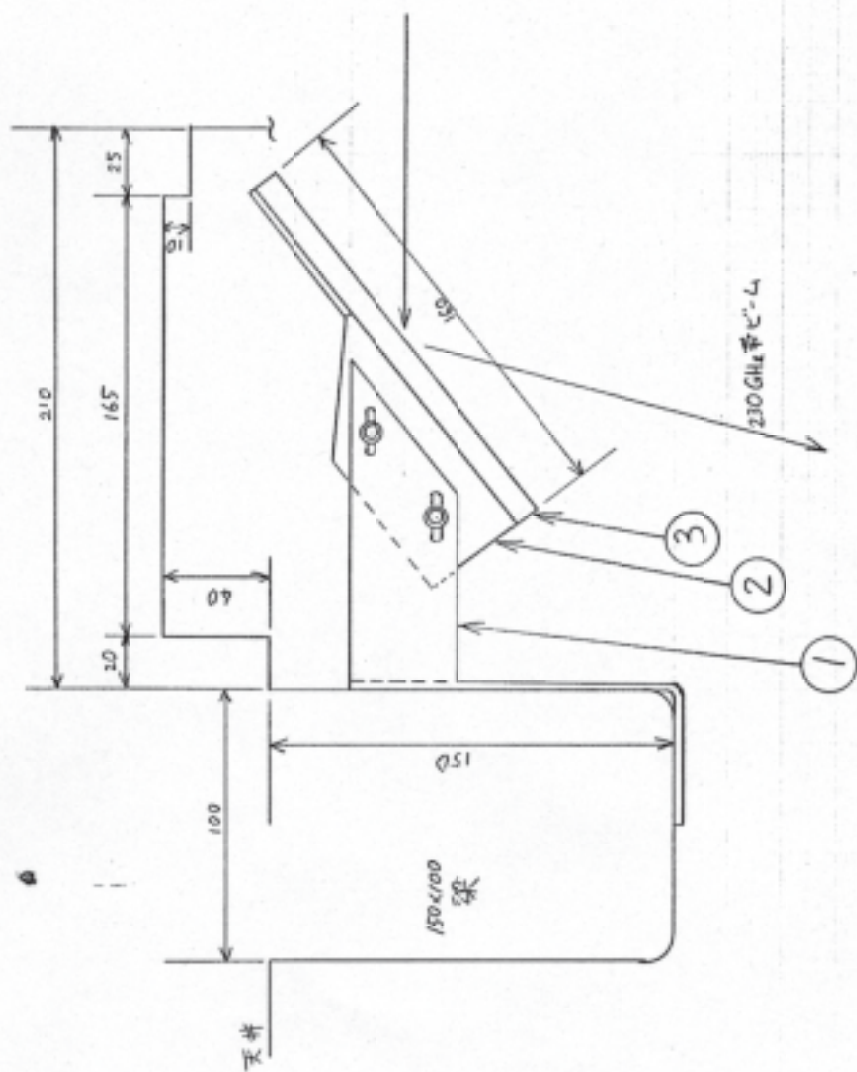
ワインを買いましたので みな様でめし上げて下さい。

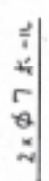
おまけ HOT/COLDの設計図(実物とイカうとに7枚ある)を一部おいておきます 落合
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184 東京都小金井市栗井北町4-2-1

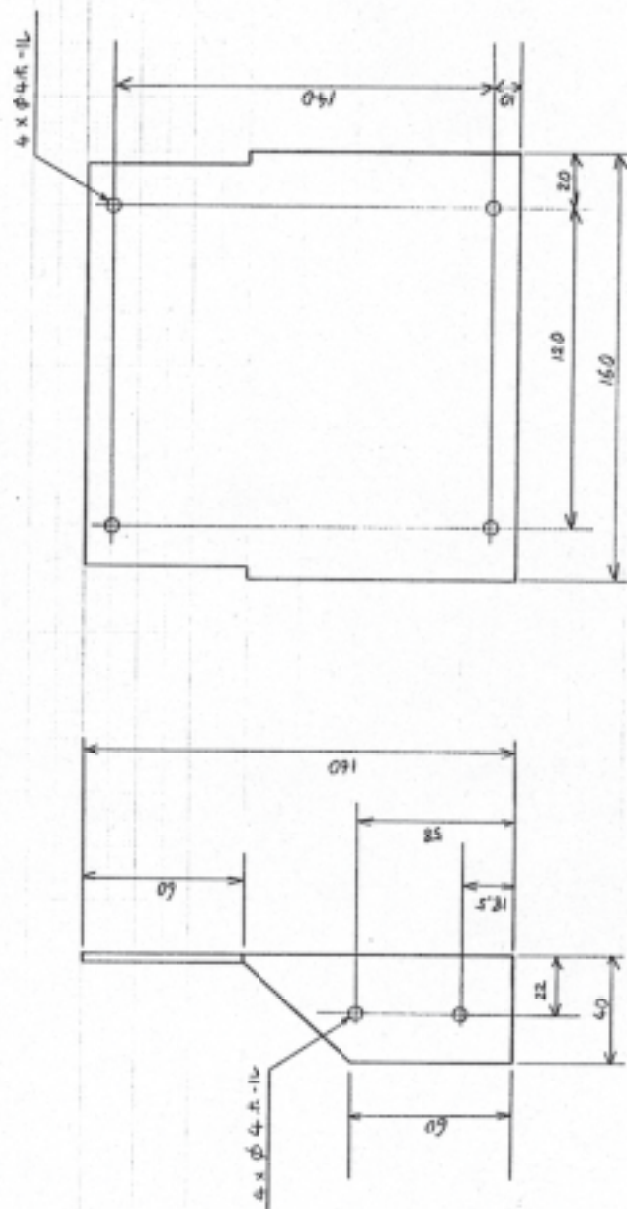
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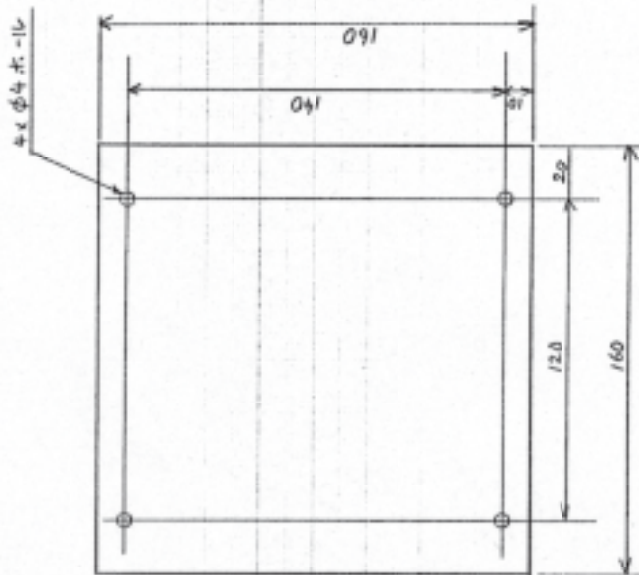


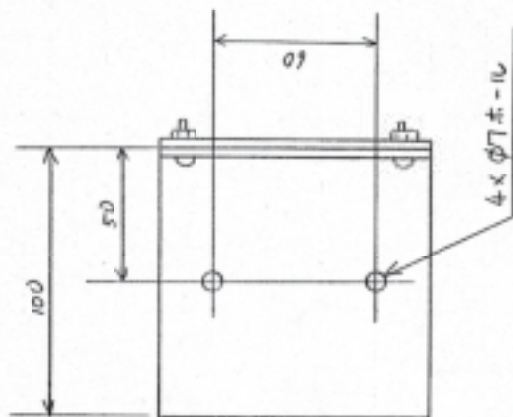
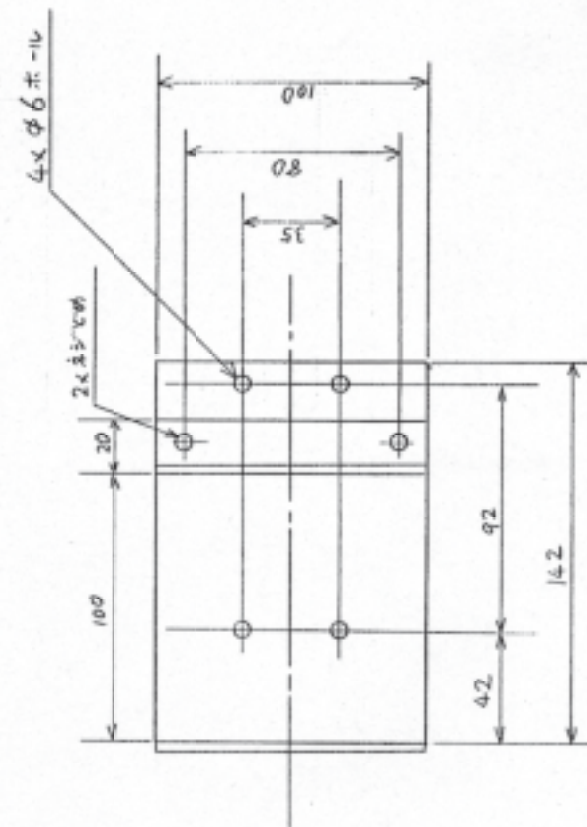
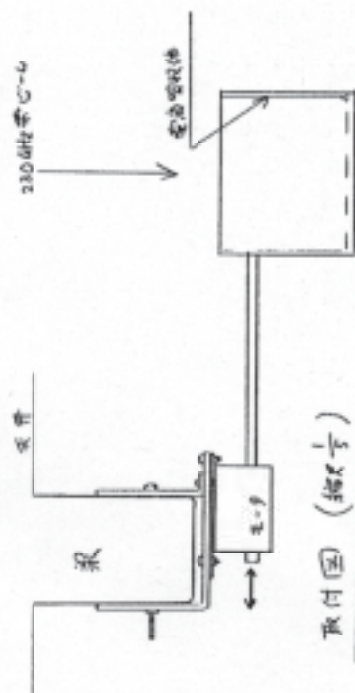
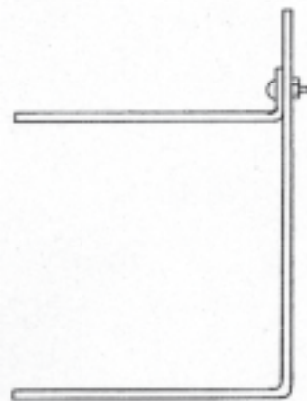


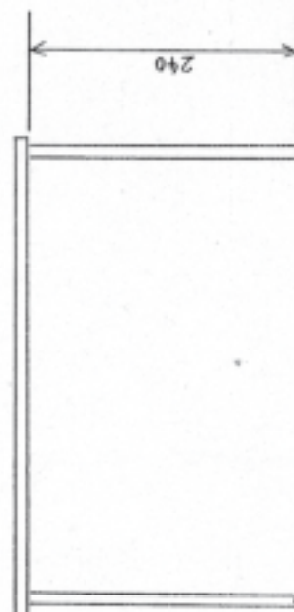
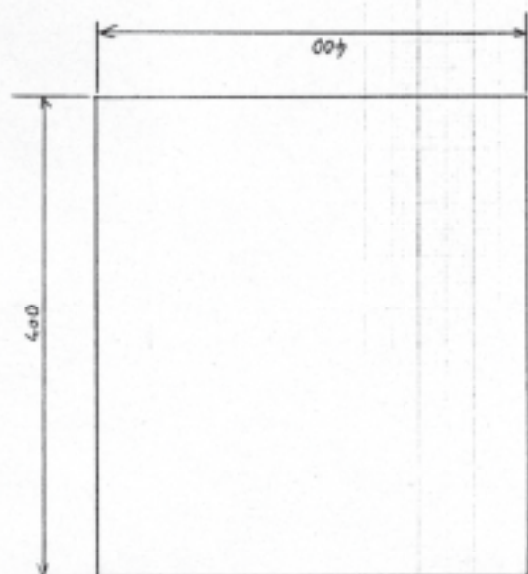
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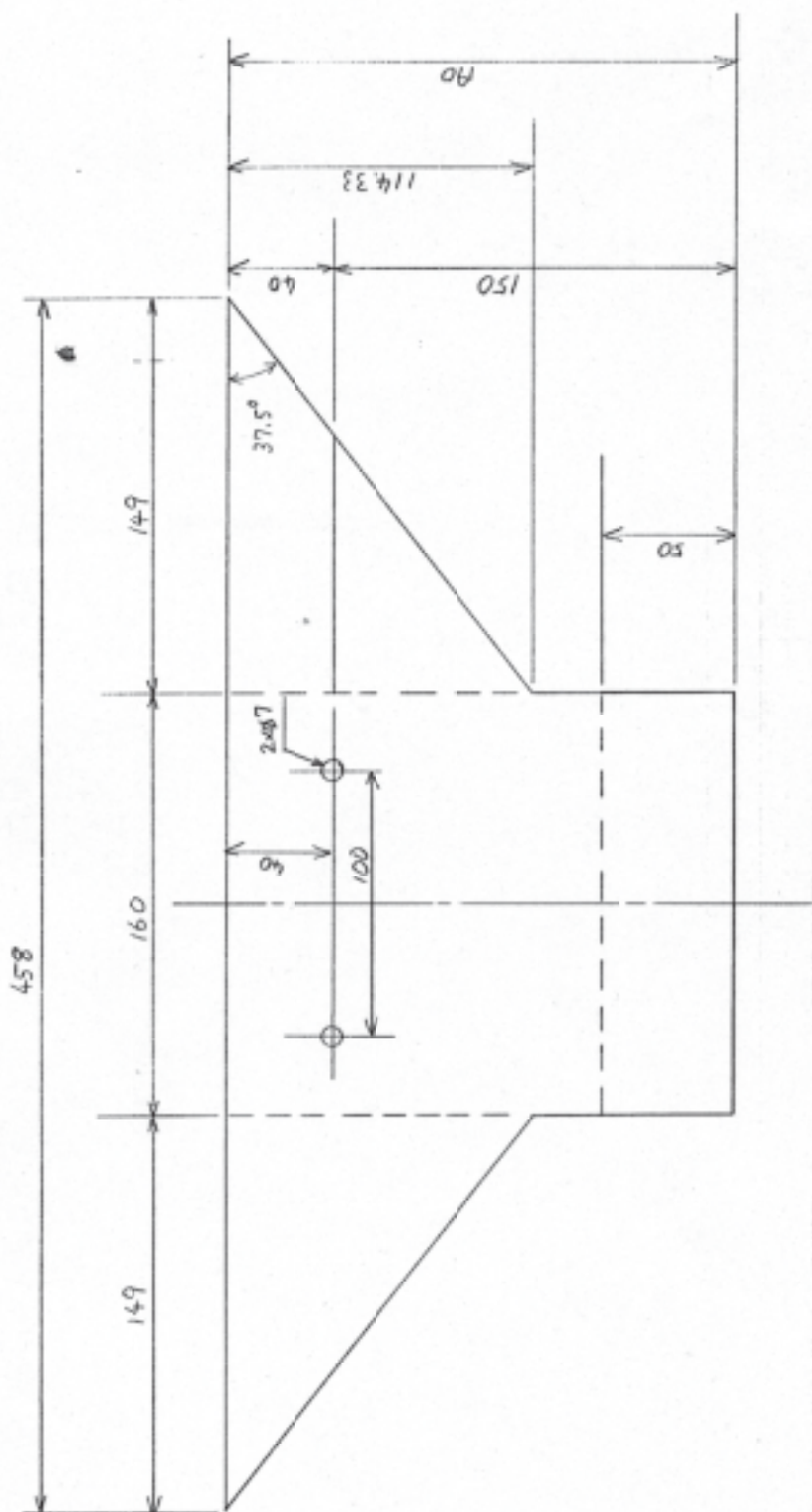


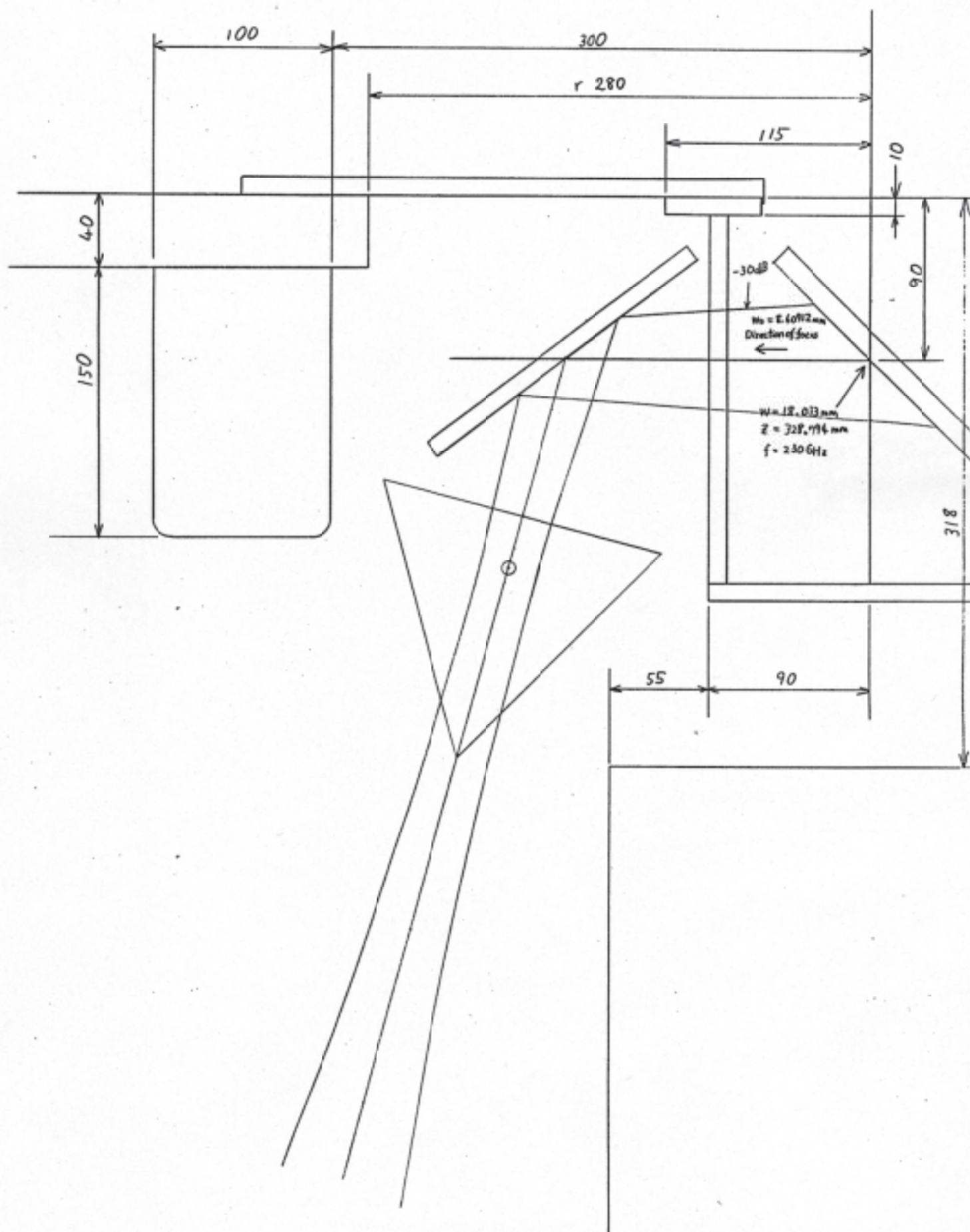
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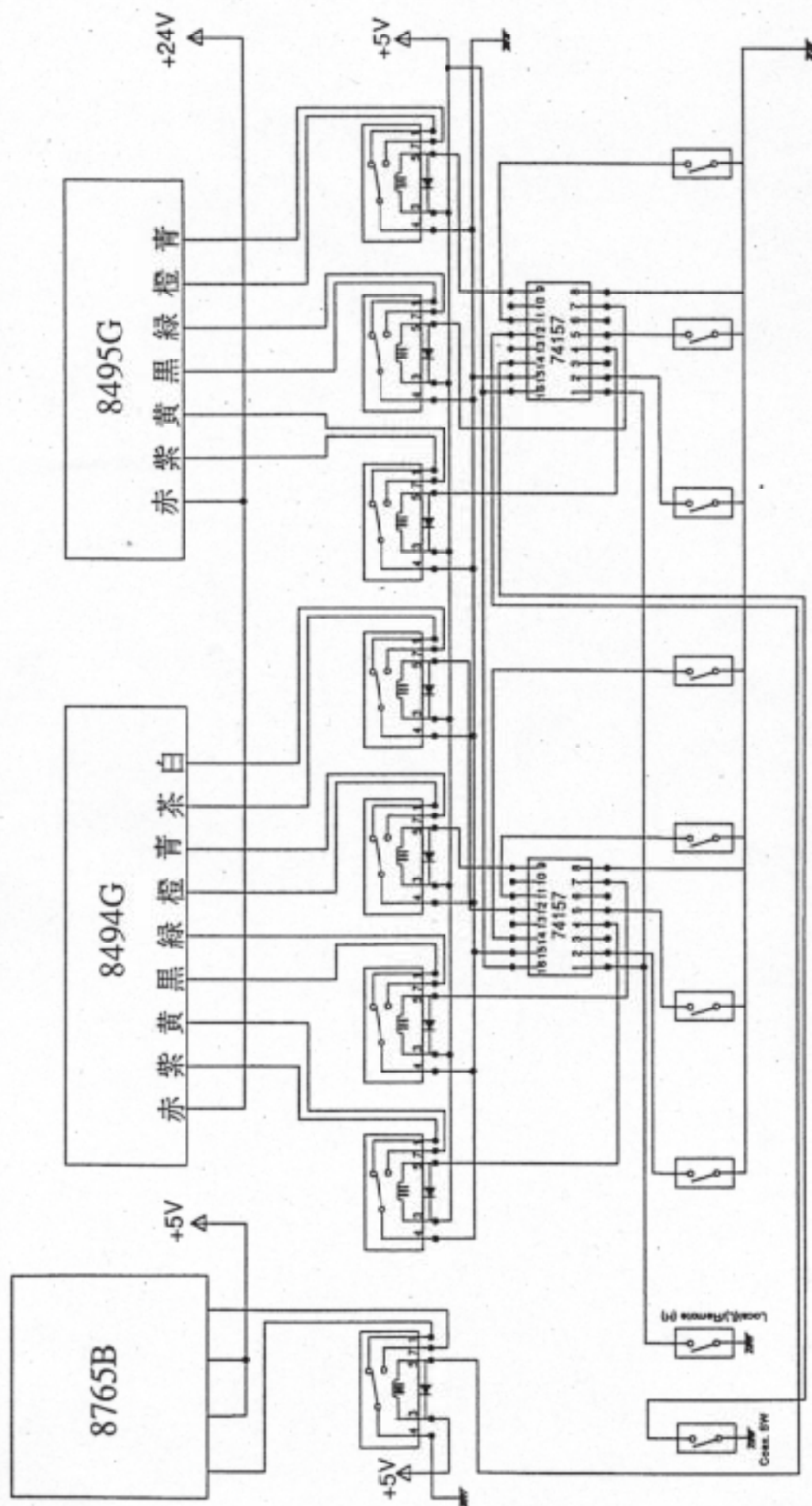




東大 60cm 望遠鏡 HOT/COLD LOAD SW 付近

線径 $\frac{1}{2}$ (13n \pm)

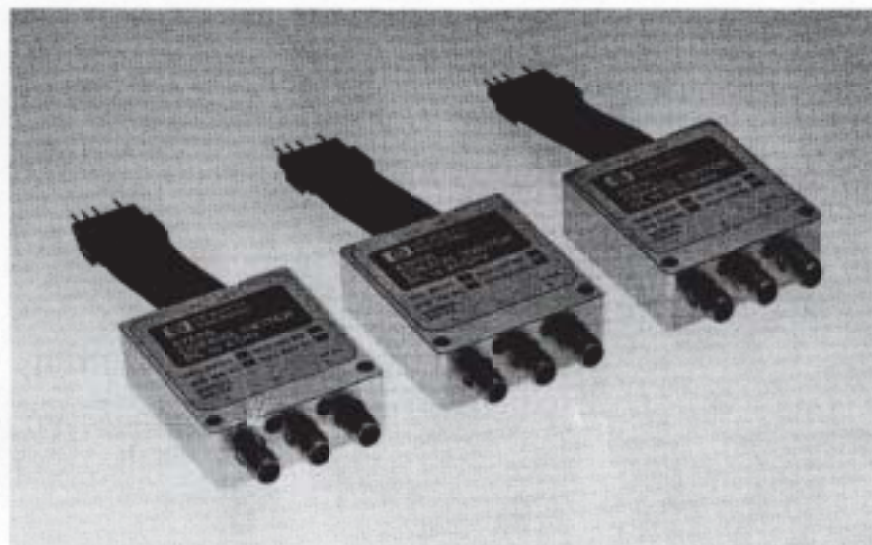
1997.2.7
高倉



HP 8765A/B/C Microwave SPDT Switches

DC - 4, 20, 26.5 GHz

Technical Data



Features and Description

- High Isolation
- Exceptional reliability, long life
- Excellent repeatability
- Unterminated

The HP 8765 family of switches is designed for microwave instrumentation and ATE systems use and features excellent electrical and mechanical performance.

The HP 8765A comes with SMA connectors for use from DC to 4 GHz, the HP 8765B also has SMA connectors and covers DC to 20 GHz, and the HP 33314C has 3.5 mm connectors and covers the DC to 26.5 GHz range. All the switches are available with voltage options covering any drive voltage between 4.5 volts up to 30 volts DC. Optional performance data is available as Option 890.

This family of switches was designed from the ground up for maximum dependability. The design goals included lowering cost and increasing the life. To do this, it was necessary to take full advantage of Hewlett-Packard's capabilities in machining and plating processes.

Long Life and Elegance

In electromechanical switches, to achieve long life, it was necessary to reduce or eliminate as many causes of failures as possible. Among the possible failure modes are structural failure of switching components, failure of the magnetic circuit, or degradation of microwave performance through debris collection or corrosion on internal microwave contacts.

The HP 8765 family of switches uses only four moving parts to accomplish the switching action. This means that there are fewer parts that can break, and fewer surfaces to generate frictional debris. The parts that do move are the rocker arm, which is the link between the magnetic circuit and the sliders, the 2 sliders, which move the edge-line jumper to form a transmission line between two of the three connectors at a time, and the return spring, which provides the linkage between the sliders.

The magnetic circuit was designed to work through a wide switching voltage range to give the user a more versatile and robust switch. The rocker arm, which is controlled by a coil and permanent magnet circuit, never comes in contact with the coils. This eliminates the possibility of impact damage and debris generation.

The sliders used to move the edge-line center conductor to the connectors to complete the microwave path are precision molded in Hewlett-Packard's plastic molding facility. Control over the dimensions of the finished molded part is critical to the reliability of this switch family. In particular, these sliders have guide channels in them which work with precisely ground and positioned dowel pins to precisely control the motion of the sliders in two ways. The first minimizes misalignment and torqueing of the sliders as they move, keeping

Driving the HP 8765A/B/C

them perpendicular to the plane of the RF contacts. The second dimension controlled is the distance the sliders can travel. This is very important, as it will determine the contact pressure between the jumpers and the RF contacts. The contact pressure between the jumper and RF contact is critical to the long life and repeatability of the switch. Too much pressure will damage both the jumper and contact, too little will not provide enough for a repeatable electrical contact. The RF contacts on the edge-line jumpers and the RF connector's center conductors are plated in Hewlett-Packard's plating facility to assure plating quality.

The spring, which links the two sliders, has been designed to provide consistent pull back pressure on the unselected jumper/slider combination. It is also designed to work in conjunction with the alignment dowel pins to prevent any unwanted torque on the sliders as they are moved.

Finally, the bodies of the switches and the moving parts have been selected and designed to provide sustained trouble free life. The surface finishes of the internal portions of the body which come into contact with the sliders, springs and rocker arm are machined to a smooth finish and then plated to minimize wear. This helps prolong life in two ways, first by reducing the degradation of materials by removing material and second by reducing the generation of wear debris. Wear debris can eventually find its way between the jumpers and RF contacts, increasing Insertion Loss and decreases switch repeatability. By controlling the amount of wear debris, and providing a wiping action of sufficient force between the jumpers and contacts, repeatability is maintained.

The HP 8765 Switch family is available with two methods for hooking up to the dc control circuitry. The standard switch comes with a ribbon cable terminated with a single in-line five position male connector with one pin removed. The user has the option of ordering solder terminals (Option 100) where they are required. (Figure 3)

The HP 8765 Switches have both sides of the switching solenoids available to the user. Typically, as is the case with the HP 8762 switches, the switching solenoids will have a common positive terminal or a common ground. By making both sides of each coil available, users can select the dc drive method that best suits their needs. The suggested methods are:

- 1) Common Negative Drive
- 2) Common Positive Drive
- 3) Polarity Reversal Drive

Regardless of which method is chosen for switching, two conditions will always apply: 1) the switch is positive latching and 2) since there are no dc current interrupts, the supply may be continuous or may be switched off after the 20 mSec switching time.

Common Negative Drive

With the common negative drive method, the two outside pins on the ribbon cable or the two outside terminals on the solder block are connected together to form a common negative or ground. Switching is then accomplished by applying the appropriate positive voltage to either of the inner pins or terminals. (Figure 1)

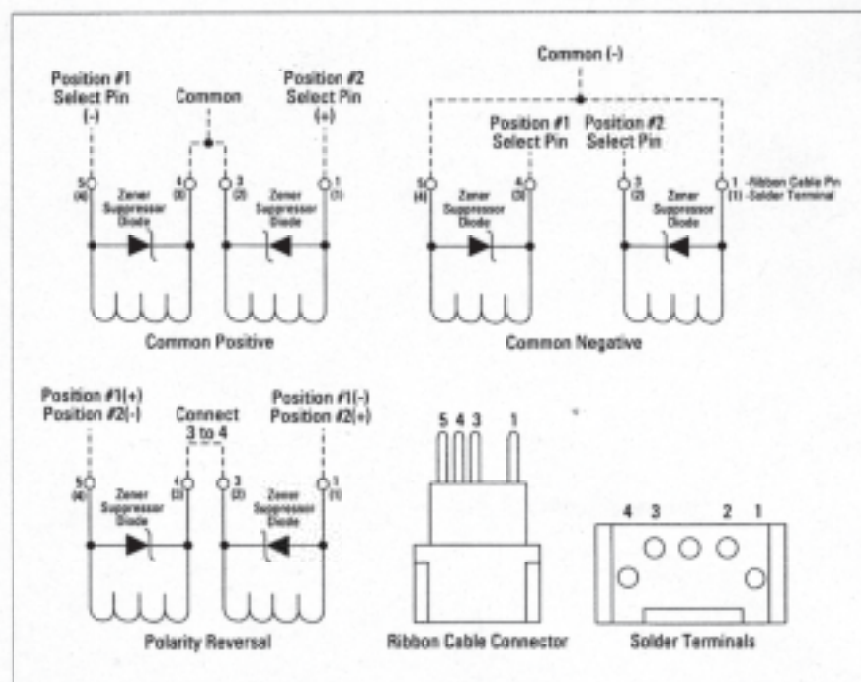


Figure 1.
DC Control Circuit
Configurations.

Common Positive Drive

To drive the switch with a common positive arrangement, the two inner pins or terminals are tied together and then to a positive voltage. To affect switching, one of the two outer pins or terminals is brought to ground. (Figure 1)

Polarity Reversal Drive

To drive the switch using a polarity reversal scheme, the two inner pins or terminals are tied together. The switch will complete a circuit between the two RF contacts on the side of the switch to which the negative or ground potential has been connected. (Figure 1)

The single in-line male connector will accommodate any receptacle which mates to 0.64mm (0.025 inch) square pins on 2.54mm (0.100 inch) centers (for instance the DuPont Berg Mini-PV receptacle). Options 108 and 116 add an 8 inch or 16 inch extension cable. Option 100 specifies solder terminals in the place of the ribbon cable.

Easy HP-IB Programmability

The HP 8765 switch family can be easily used in ATE systems with HP-IB control using the HP 11713A Attenuator/Switch Driver. Connections for drive control are easily made by ordering the HP 11761A cable/adaptor, which plugs into the HP 11713A and will connect up to four HP 8765A/B/C switches. Interconnecting between the HP 11713A and the HP 8765A/B/C switches with solder terminals (Option 100) is possible with the standard cables supplied with each HP 11713A. Each HP 8765 Option 024 continuously draws 120 mA, due to this constant draw of power, the HP 11713A can only accommodate five HP 8765's at a time.

Specifications

HP Model:	HP 8765A	HP 8765B	HP 8765C
Frequency range:	dc-4 GHz	dc-20 GHz	dc-26.5 GHz
Insertion loss (dB): (f is frequency in GHz)	$0.2 + 0.025f$	$0.2 + 0.025f$	$0.25 + 0.027f$
*Typical insertion loss:	0.2 @ 4 GHz	0.2 @ 4 GHz 0.5 @ 20 GHz	0.2 @ 4 GHz 0.5 @ 20 GHz 0.7 @ 26.5 GHz
Isolation (dB): (f is frequency in GHz)	120-2.25f	120-2.25f	120-2.6f
*Typical isolation:	120 @ 4 GHz	120 @ 4 GHz 90 @ 20 GHz	120 @ 4 GHz 90 @ 20 GHz 60 @ 26.5 GHz
SWR:	<1.2 dc-4 GHz	<1.2 dc-4 GHz <1.35 4-12.4 GHz <1.45 12.4-18 GHz <1.7 18-20 GHz	<1.25 dc-4 GHz <1.45 4-18 GHz <1.7 18-26.5 GHz
Connectors:	SMA (f)	SMA (f)	3.5 mm(f)
Repeatability: (Up to 5,000,000 Cycles measured at 25° C, dc)	<0.03 dB	<0.03 dB	<0.03 dB

*Specification describe the instrument's warranted performance. Supplemental and typical characteristics are intended to provide information useful in applying the instrument by giving typical, but not warranted performance parameters.

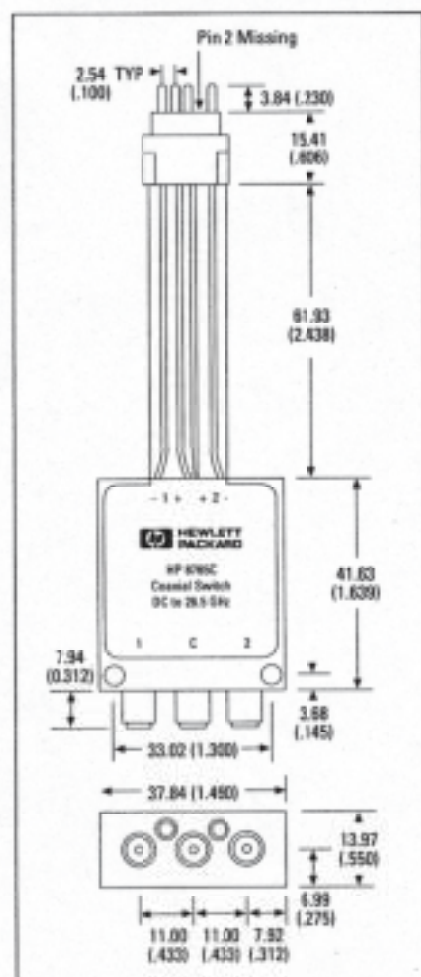


Figure 2. Standard unit part dimensions in millimeters and (inches).

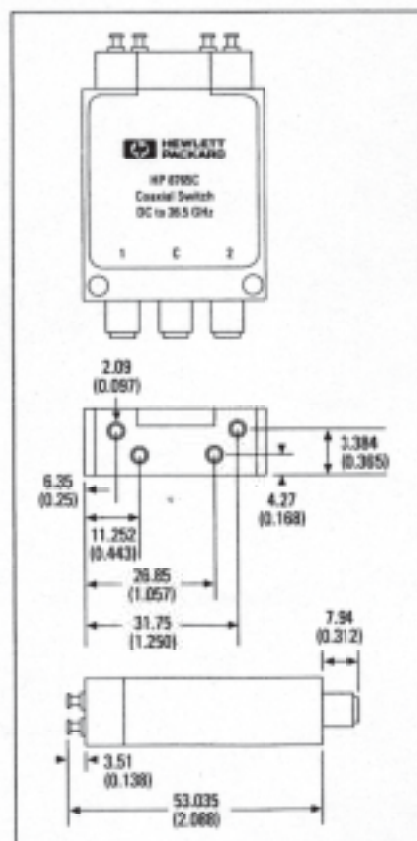


Figure 3. Option 100 with solder terminals part dimensions in millimeters and (inches).

General Operating Data



Maximum Power Rating: 2 watt average for switching with power applied.
100 watt peak, non-switching

Life: > 5,000,000 Cycles

Switching speed: 15msec maximum

Solenoid Electrical Specifications and Switching Voltage Options

Options	Voltage	Current	Impedance
005	5 (4.5–7) Vdc	385 mA @ 5 Vdc	13 ohm, 8 mH
010	10 (7–12) Vdc	300 mA @ 10 Vdc	33 ohm, 25 mH
015	15 (12–20) Vdc	200 mA @ 15 Vdc	75 ohm, 55 mH
024	24 (20–30) Vdc	120 mA @ 24 Vdc	200 ohm, 135 mH

HP Model Calibration Frequencies

8765A	200 to 400 MHz every 200 MHz
8765B	200 to 18000 MHz every 200 MHz
8765C	1500 to 26500 MHz every 250 MHz

Environmental

Operating Temperature:

-25° to 75°C

Storage Temperature:

-55° to 85°C

Temperature Cycling:

-55° to 85°C, 10 cycles per
MIL-STD-883 method 1010

Vibration:

Operating: 7 g's; 5-2000 Hz @

0.25" p-p

Survival: 20 g's; 20-2000 Hz @

0.06" p-p 4 min/cycle, 4 cycles/
per axis MIL-STD-883
method 2007

Random: Survival: 2.41 g(rms)
10 min/axis

Shock:

Half Sine: 500 g's @ 0.5 mS,
3 drops/direction, 18 total per
MIL-STD-883 method 2002.1

Operating: 50 g's @ 6 mS,
6 directions

Moisture Resistance: 65°C,
95% RH, 10 days per MIL-STD-
883 method 1004

Altitude Storage: 50,000 ft.
(15,240 meters) per MIL-STD-
883 method 1001, for > 1 hour

RFI: per MIL-STD-461C, RE02,
part 4

Electrostatic Discharge:

25 kV max

Abuse: 2W, CW, 1 kW peak pulse
(10 μ s), 2 minute duration

Operating Life:

>5,000,000 cycles

Ordering

Information

Specify switching voltage option
(005, 010, 015, or 024).

Specify flexible cable extensions if
needed:

Option 108: 8 inch (20.3 cm)

Option 116: 16 inch (40.6 cm)

Option 100 solder terminals are
available in place of ribbon drive
cable.

HP 11761A Cable/Adapter: this 5-
foot cable connects directly be-
tween the HP 11713A Attenuator/
Switch Driver and the ribbon
cable of up to four HP 8765A/B/C
switches.

Option 890 SWR and insertion
loss data measured with an auto-
matic network analyzer with very
small uncertainties; directly
traceable to NIST standards.

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May 1989

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5953-2351

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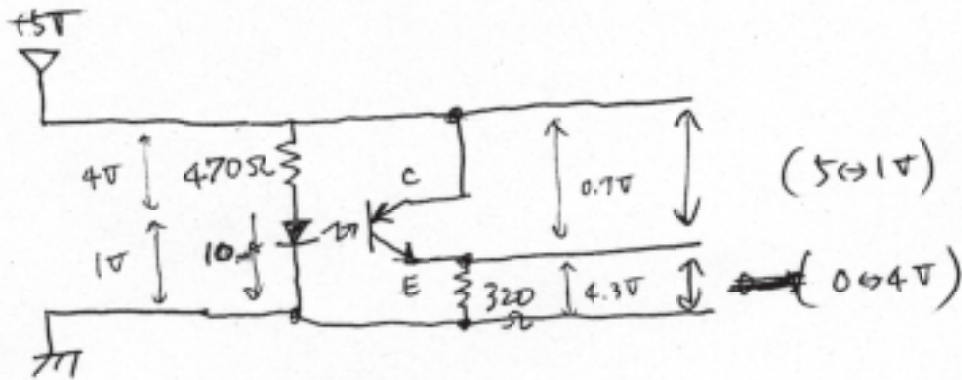
フォトインタラプターの使い方



TP507A



← 裏向き



向別の数を数えることができます。また、種々の制御に使うこともできます。

フォト・センサは発光、受光間の距離、条件によって定常光を用いるか、あるいは変調光を用いるかを決めればOKです。

⑤ フォト・カプトラによるインタフーズ回路

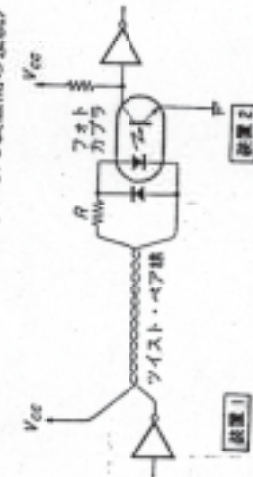
フォト・カプトラとして最も有効で、かつ多く用いられる個所として、二つの装置のインタフーズがあります。

装置と装置、たとえばコンピュータ・システムと端末機器といった装置を結ぶ場合、接地ループやそれに関連する誘導、雑音などの問題で誤動作を起こすことがあります。

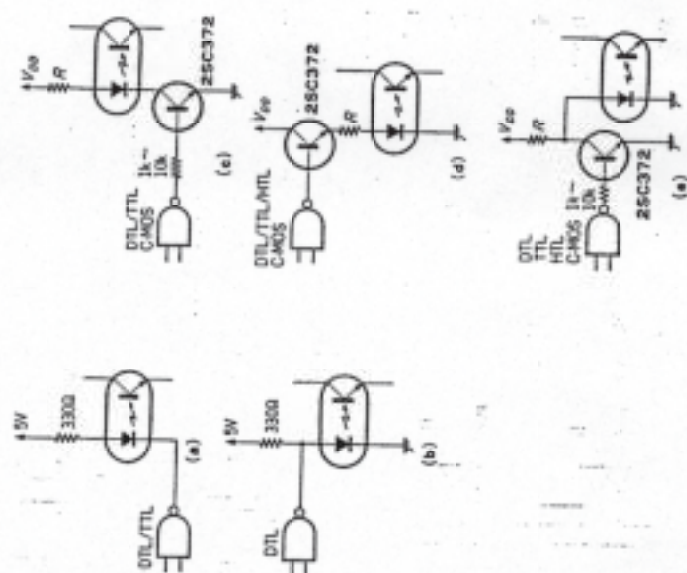
このような影響を除くために、従来、リード・リレーなどを用いていましたが、処理スピードが遅いという欠点がありました。そこで、図5-78のようにフォト・カプトラを用いることにより、両装置を電気的に絶縁することができ、接地ループなどの問題を解決することができます。

このようにフォト・カプトラは電気的分極に使われますが、その駆動方法にはいろいろな方法があります。図5-79はいくつかのデジタルICによるフォト・カプトラの駆動回路を示します。DTLやTTLで駆動するときはそのまま接続が可能です。が、C-MOSで直接駆動するの

〈図5-78 フォト・カプトラによる装置間の接続〉



〈図5-79 フォト・カプトラの駆動回路〉

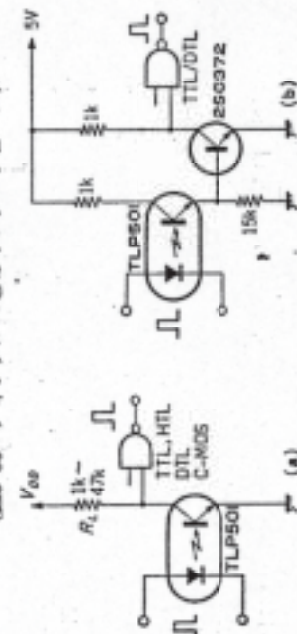


はちょっと苦しくなります。したがって、同図(c)(d)のようにトランジスタを用いるのが簡単といえます。図において、電流制限抵抗 R は LED に流す電流値に応じて選定します。

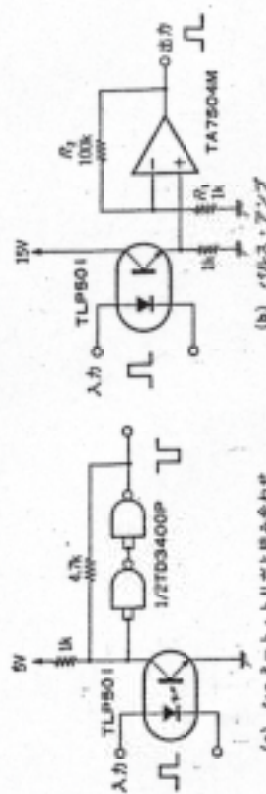
次に、フォト・トランジスタ側の回路を図5-80に示します。TLP501などは変換効率が10%以上ですからLED側の駆動電流を20~30mA流せばTTLやDTLを直接ドライブできますが、LED側電流が小さいときは、同図(b)のようにトランジスタで電流増幅を行なった後、駆動する必要があります。

また、フォト・カプトラの応答時間は数 μ s となりますが、それを整形するには、図5-81(a)のようにシュミット回路を構成すると有効です。同図(b)はOPアンプを用いた回路で、LED側の入力がいまいち時効です。OPアンプの利用は $(R_1 + R_2)/R_3$ で与えられますから、適当に選ぶことにより、入力がいまいち小さくても充分といえます。

〈図5-80 デジタルICとのインターフェース〉



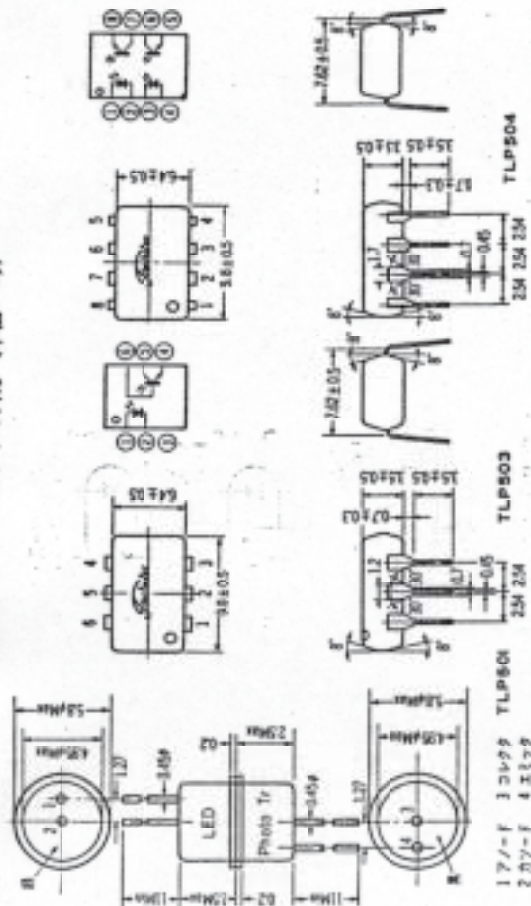
〈図5-81 入力回路〉



● フォト・カプトラによる検出回路

フォト・カプトラのおもしろい応用として、直接検出回路を組むことができます。図5-82はその回路で、たとえば、図(a)の回路では、二つの入力に“1”が入ったときだけ出力が

〈図5-28 フォト・カプラの外形 (単位mm)〉

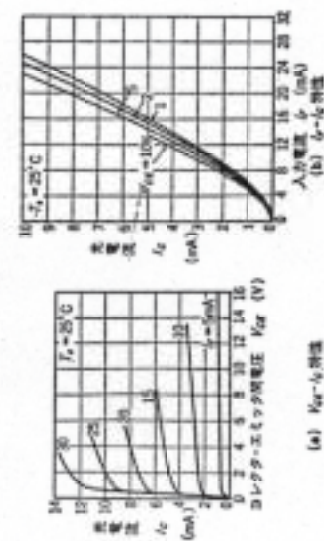
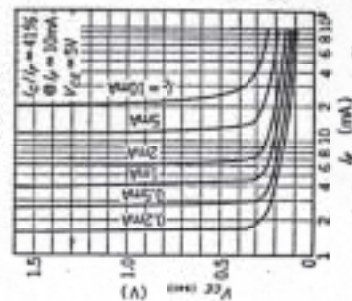


それぞれ DIP の 6 ピン、8 ピンに封入されています。したがって、TLP503、504 はディジタル IC と組み合わせ使用するとき便利となります。

フォト・カプラの重要なパラメータとして信号の伝達効率がありますが、LED の順電流 I_F に対してフォト・トランジスタのコレクタ電流 I_C の比を取り、 I_C/I_F を変換効率と呼びます。表 5-3 の特性では標準 30% ですから、 $I_F=10\text{mA}$ のとき $I_C=3\text{mA}$ ということになります。

図 5-29 は変換特性をあらわしたもので、(a) はフォト・トランジスタの $V_{CE}-I_C$ 特性で、 I_F をバ

〈図5-29 フォト・カプラの伝達特性 (TLP501)〉

〈図5-30 I_F-V_{CE} 特性 (TLP503)〉

ラメータにしたものです。(a) は変換特性を示しますが、小電流で使用するときより $I_F > 10\text{mA}$ で使用するほうが効率がよいといえます。

また、トランジスタが飽和状態にあるとき、そのコレクタ-エミッタ間の残り電圧 $V_{CE(sat)}$ は普通 0.2~0.4V 程度ですが、図 5-30 に I_C をパラメータにした I_F と $V_{CE(sat)}$ の関係を示

〈表5-3 フォト・カプラの電気的特性例 ($T_A=25^\circ\text{C}$)〉

最大定格⇒

項目	記号	定価	単位
LED		TLP503, 504	
順電流 (I_F)	I_F	60	mA
順電流 (Pulse) *	I_{FP}	1	A
直流電圧	V_F	6	V
許容損失	P_C	100	mW
伝達率	$\Delta P/\Delta T$	1.3	mW/°C
コレクタ-ベース間電圧	V_{CB}	50	V
コレクタ-エミッタ間電圧	V_{CE}	45	V
エミッタ-コレクタ間電圧	V_{EC}	6	V
エミッタ-ベース間電圧	V_{EB}	5	V
許容損失	P_C	100	mW
伝達率	$\Delta P/\Delta T$	2	mW/°C
動作温度	T_{OP}	-30~100	°C
保存温度	T_{STG}	-55~125	°C
許容損失 (503)	P_T	200	mW
伝達率 (503)	$\Delta P/\Delta T$	2.6	mW/°C
許容損失 (504)	P_T	400	mW
伝達率 (504)	$\Delta P/\Delta T$	5.3	mW/°C

0 電気的特性

項目	記号	条件	TLP501	TLP503, 504	単位
順電流電圧	V_F	$I_F=10\text{mA}$	MIN 1.35	MAX 1.25	V
逆電流	I_R	$V_F=5\text{V}$	MAX 1.0	MIN 1.0	μA
結合容量	C	$f=5\text{MHz}$	MAX 50	MIN 50	pF
コレクタ-ベース間電圧	V_{CB}	$I_C=0.5\text{mA}$	MAX 45	MIN 20	V
コレクタ-エミッタ間電圧	V_{CE}	$I_C=0.1\text{mA}$	MAX 50	MIN 50	V
エミッタ-コレクタ間電圧	V_{EC}	$I_E=0.1\text{mA}$	MAX 6	MIN 5	V
順電流	I_C	$I_F=0, V_{CE}=10\text{V}, TLP501: 30\text{V}$	MAX 500	MIN 100	mA
結合容量	C	$f=5\text{MHz}$	MAX 7	MIN 10	pF
変換効率	I_C/I_F	$I_F=10\text{mA}, V_{CE}=5\text{V}, V_F=0$	MAX 30	MIN 30	%
入力ノイズ電圧	E_n	$f=1\text{MHz}, R_L=40-60\Omega$	MAX 3	MIN 3	μV
絶縁抵抗	R_g	$V_F=100\text{V}$	MAX 10	MIN 10	$10^9\Omega$
絶縁耐圧	BV_F	DC, 1 minute	MAX 5000	MIN 5000	V
立ち上がり・立ち下り時間	$t_{r, f}$	$V_{CE}=10\text{V}, R_L=100\Omega, I_C=1\text{mA}$	MAX 5	MIN 5	μs